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In re International Application of

Koji KIKUSHIMA

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For: OPTICAL SIGNAL TRANSMITTER AND OPTICAL SIGNAL  
TRANSMISSION SYSTEM

VERIFICATION OF TRANSLATION

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APPLICATION FOR UNITED STATES LETTERS PATENT

INVENTOR(S): Koji KIKUSHIMA

INVENTION: OPTICAL SIGNAL TRANSMITTER AND  
OPTICAL SIGNAL TRANSMISSION  
SYSTEM

S P E C I F I C A T I O N

DESCRIPTION  
OPTICAL SIGNAL TRANSMITTER AND OPTICAL SIGNAL  
TRANSMISSION SYSTEM

5 TECHNICAL FIELD

[0001] The present invention relates to an optical signal transmitter used for optical transmission of wideband signals, and relates to an optical signal transmission system using this optical signal transmitter.

10 More particularly, the present invention relates to an optical signal transmitter used for optical transmission of multichannel video signals that have undergone frequency-division multiplexing and that have undergone amplitude modulation (abbreviated as "AM") or quadrature

15 amplitude modulation (abbreviated as "QAM"), and relates to an optical signal transmission system using this optical signal transmitter.

BACKGROUND ART

20 [0002] Conventionally, an optical signal transmitter and an optical signal transmission system employing a method for subjecting video signals, which have undergone frequency-division multiplexing, to frequency modulation as a single unit (this method will be hereinafter referred

25 to as an "FM batch conversion method") are known as an optical signal transmitter and an optical signal transmission system used for optical transmission of multichannel video

signals that have undergone frequency-division multiplexing and that have undergone amplitude modulation or quadrature amplitude modulation.

[0003] An optical signal transmitter and an optical signal transmission system that employ this FM batch conversion method are disclosed in Non-patent Document 1.

[0004] Fig. 1 shows a structure of a conventional optical signal transmitter and a conventional optical signal transmission system that employ the FM batch conversion method. Figs. 2A, 2B, and 2C show signal forms at point "A," point "B," and point "C" of Fig. 1, respectively. The optical signal transmission system of Fig. 1 is comprises an optical signal transmitter 80 including an FM batch conversion circuit 81, a light source 82, and an optical amplification circuit 83, an optical transmission path 85, an optical signal receiver 90 including a photoelectric conversion circuit 91 and an FM demodulation circuit 92, a set-top box 93, and a television receiver 94. Signal spectra at point "A," point "B," and point "C" of Fig. 1 are shown in Figs. 2A, 2B, and 2C, respectively. The same applies to point "A," point "B," and point "C" of each figure shown below.

[0005] In the optical signal transmitter 80 of Fig. 1, frequency-multiplexed video signals shown in Fig. 2A are converted into one wideband frequency-modulated signal shown in Fig. 2B by the FM batch conversion circuit 81. The frequency-modulated signal is subjected to intensity

modulation by the light source 82, and is further subjected to optical amplification by the optical amplification circuit 83, and is transmitted to the optical transmission path 85. In the optical signal receiver 90, the 5 frequency-modulated signal that has undergone intensity modulation is photoelectrically converted by the photoelectric conversion circuit 91, and is returned to an electric signal. This electric signal, which is a wideband frequency-modulated signal, is subjected to 10 frequency demodulation by the FM demodulation circuit 92, and the frequency-multiplexed video signals are demodulated as shown in Fig. 2C. The demodulated video signals pass through the set-top box 93, and reach the television receiver 94, whereby a desired video channel is selected.

15 [0006] Fig. 3 shows the structure of an FM batch conversion circuit that is applicable to the FM batch conversion method (see Patent Document 1, Non-patent Document 2, Non-patent Document 3, for example). The FM batch conversion circuit shown in Fig. 3 uses an optical 20 frequency modulation portion and an optical frequency local oscillation portion. The FM batch conversion circuit 81 comprises the optical frequency modulation portion 71, the optical frequency local oscillation portion 72, an optical multiplexer 73, and a photodiode 74.

25 [0007] When frequency modulation is performed with a frequency  $f_s$  by use of a carrier light source having an optical frequency  $f_o$  in the optical frequency modulation

portion 71 of the FM batch conversion circuit 81, an optical frequency  $F_{fmld}$  of an optical signal in the output of the optical frequency modulation portion 71 is expressed as in the following equation:

5  $F_{fmld} = f_0 + \delta f \cdot \sin(2\pi \cdot f_s \cdot t) \quad (1)$

where  $\delta f$  is a frequency deviation. A DFB-LD (Distributed Feed-Back Laser Diode) is used as the carrier light source of the optical frequency modulation portion 71.

[0008] In the optical frequency local oscillation portion 72, oscillation is performed by use of an oscillation light source having an optical frequency  $f_1$ . An optical signal transmitted from the local oscillation portion 72 and an optical signal transmitted from the optical frequency modulation portion 71 are multiplexed by the optical multiplexer 73. The DFB-LD is used as the oscillation light source of the optical frequency local oscillation portion 72. The two optical signals multiplexed by the optical multiplexer 73 are detected by the photodiode 74 that is an optical heterodyne detector. The frequency  $f$  of the electric signal detected thereby is expressed as follows:

20  $f = f_0 - f_1 + \delta f \cdot \sin(2\pi \cdot f_s \cdot t) \quad (2)$

Herein, if the optical frequency of the carrier light source of the optical frequency modulation portion 71 and the optical frequency of the oscillation light source of the optical frequency local oscillation portion 72 are caused to come close to each other, it is possible to obtain an electric signal whose frequency is modulated to have an

intermediate frequency  $f_i = f_o - f_1$  of several GHz and have a frequency deviation  $\delta f$  as shown in Fig. 2B.

[0009] Generally, the modulation by an input electric current allows the DFB-LD to have an optical frequency varied 5 in the range of several GHz in accordance with the input electric current, and hence a value of several GHz can be obtained as the frequency deviation  $\delta f$ . For example, a multichannel AM video signal or QAM video signal that have undergone frequency multiplication so as to have a frequency 10 range of about 90 MHz to about 750 MHz can be converted by the FM batch conversion circuit into a frequency-modulated signal having a frequency band of about 6 GHz in which the intermediate frequency  $f_i = f_o - f_1$  becomes equal to about 3GHz as shown in Fig. 2B.

[0010] Fig. 4 shows the structure of an FM demodulation 15 circuit applicable to the optical signal receiver 90. The FM demodulation circuit 92 shown in Fig. 4 is an FM demodulation circuit by delay-line detection, and comprises a limiter amplifier 76, a delay line 77, an AND gate 78, 20 and a low-pass filter 79.

[0011] In the FM demodulation circuit 92, a frequency-modulated optical signal that has been input is shaped into a square wave by the limiter amplifier 76. The output of the limiter amplifier 76 is branched into two 25 output parts, one of which is input to an input terminal of the AND gate 78 and the other of which undergoes a polarity reversal, is then delayed by time  $t$  by means of the delay

line 77, and is input to an input terminal of the AND gate 78. The output of the AND gate 78 is smoothed by the low-pass filter 79, and is turned into frequency-demodulated output (see Non-patent Document 1, for example).

5 [0012] A double-tuned frequency discriminator having a resonance circuit, a Foster-Seeley frequency discriminator, and a ratio detection type FM demodulator can be mentioned as a circuit form of the FM demodulation circuit, in addition to the FM demodulation circuit by 10 delay-line detection described here.

[0013] Patent Document 1: Japanese Patent No. 2700622; Non-patent Document 1: international standard, ITU-T J. 185, "Transmission equipment for transferring multi-channel television signals over optical access 15 networks by FM conversion;"

Non-patent Document 2: Shibata et al. "Optical image distribution system using an FM batch conversion method," Institute of Electronics, Information and Communication Engineers, Technical Journal B, Vol. J83-B, No. 7, July, 20 2000, pp. 948-959;

Non-patent Document 3: Suzuki et al. "Pulsed FM batch conversion modulation analog optical CATV distribution method" Institute of Electronics, Information and Communication Engineers, Autumn Conference, B-603, 1991.

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#### DISCLOSURE OF THE INVENTION

[0014] A low noise and a low distortion are required

in transmission of the multichannel video signals described above. According to "Optical image distribution system using an FM batch conversion method" by Shibata et al., a CNR (Carrier-to-Noise Ratio) is set to be 42 dB or more, and a CSO (Composite Second-Order Distortion) and a CTB (Composite Triple Beat) are set to be -54 dB or less in an optical signal transmitter and an optical signal transmission system using an FM batch conversion method.

[0015] However, in the optical signal transmitter using the conventional FM batch conversion method, the CNR value is in a saturated state between 43 dB to 47 dB. Likewise, the CSO value and the CTB value are in a saturated state having a value slightly below -54 dB. If the optical signal transmitter can be constructed to have an even lower noise, the CNR can be enlarged, and, as a result, the minimum electric power of the optical signal receiver whose CNR is 42 dB or more can be reduced. If the minimum light-receiving power of the optical signal receiver can be reduced, the transmission distance can be lengthened, and the optical branching ratio can be enlarged.

[0016] The DFB-LD of the optical frequency modulation portion used in the conventional FM batch conversion circuit proves difficult in modification of its design when returning to its structure, and it was difficult to realize low-noise characteristics and low-distortion characteristics. It is therefore an object of the present invention to provide an optical signal transmitter low in

noise and in distortion and provide an optical signal transmission system using this optical signal transmitter.

[0017] In order to achieve this object, according to a first aspect of the present invention, the present invention is characterized in that an optical signal transmitter for applying frequency modulation to amplitude-modulated electric signals that have undergone frequency division multiplexing to optically transmit the electric signals, the optical signal transmitter comprising: a distribution circuit for distributing the electric signals into a plurality of signal parts and outputting the signal parts; a plurality of frequency modulation means for applying frequency modulation to each output of the distribution circuit and emitting each output, the plurality of frequency modulation means being substantially equal to each other in frequency deviation and in intermediate frequency and being substantially identical in the phase of each output; a multiplexing means for multiplexing outputs of the plurality of frequency modulation means and outputting multiplexed outputs; and a transmitting circuit for outputting optical signals subjected to intensity modulation by the output of the multiplexing means to an optical transmission path. Herein, the electric signals that have undergone frequency-division multiplexing and amplitude modulation include electric signals that have undergone frequency-division multiplexing and quadrature amplitude modulation.

[0018] According to a second aspect of the present invention, the present invention is characterized in that an optical signal transmission system comprises the optical signal transmitter according to the first aspect of the present invention, a photoelectric conversion means connected to the optical signal transmitter through an optical transmission path, and an optical signal receiver having a frequency demodulation means for applying frequency demodulation to an output of the photoelectric conversion means.

[0019] The optical signal transmitter and the optical signal transmission system according to the present invention can obtain lower noise characteristics and lower distortion characteristics than a conventional optical signal transmitter while using conventional electric circuits and conventional optical circuit components without changing circuit constants returning to the circuit design of such electric circuits and optical circuit components.

[0020] The low noise characteristics of the optical signal transmitter make it possible to reduce the minimum light-receiving electric power of the optical signal receiver, thus making it possible to lengthen the transmission distance and to enlarge the optical branching ratio between the optical signal transmitter and the optical signal receiver.

[0021] Additionally, the low distortion

characteristics thereof make it possible to improve a video-signal receiving quality.

#### BRIEF DESCRIPTION OF THE DRAWINGS

5 [0022] Fig. 1 is a block diagram showing a structure of a conventional optical signal transmitter and a conventional optical signal transmission system using an FM batch conversion method;

10 Fig. 2A is a view showing signal forms in the optical signal transmitter and the optical signal transmission system;

15 Fig. 2B is a view showing a signal form in the optical signal transmitter and the optical signal transmission system;

20 Fig. 2C is a view showing signal forms in the optical signal transmitter and the optical signal transmission system;

25 Fig. 3 is a block diagram showing a structure of a conventional FM batch conversion circuit applicable to the FM batch conversion method;

30 Fig. 4 is a block diagram showing a structure of an FM demodulation circuit applicable to an optical signal receiver;

35 Fig. 5 is a block diagram showing a structure of an optical signal transmitter in which N FM batch conversion circuits, to which electric signals distributed by a distribution circuit are input while being modulated, are

used;

Fig. 6 is a block diagram showing a structure of an FM batch conversion circuit that is applied to an optical signal transmitter and that uses an optical frequency modulation portion;

Fig. 7 is a block diagram showing a structure of an FM batch conversion circuit that is applied to an optical signal transmitter and that uses two optical frequency modulation portions for a push-pull structure;

Fig. 8 is a block diagram showing a structure of an FM batch conversion circuit that is applied to an optical signal transmitter and that uses a voltage-controlled oscillator;

Fig. 9 is a block diagram showing a structure of an FM batch conversion circuit that is applied to an optical signal transmitter and that uses two voltage-controlled oscillators for a push-pull structure;

Fig. 10 is a block diagram showing a structure of an optical signal transmitter including two sets of an optical frequency modulation portion and optical frequency local oscillation portion;

Fig. 11 is a block diagram showing a structure of an optical signal transmitter in which two optical frequency modulation portions are used for a push-pull structure and in which two push-pull structures are used;

Fig. 12 is a block diagram showing a structure of an optical signal transmitter in which N optical frequency

modulation multiplexing circuits, to which electric signals distributed by a distribution circuit are input while being modulated, are used;

5 Fig. 13 is a block diagram showing a structure of the optical frequency modulation multiplexing circuit;

10 Fig. 14 is a block diagram showing a structure of an optical signal transmitter in which N differential optical frequency modulation multiplexing circuits, to which electric signals distributed by a distribution circuit are input while being modulated, are used; and

15 Fig. 15 is a block diagram showing a structure of the differential optical frequency modulation multiplexing circuit.

#### 15 BEST MODE FOR CARRYING OUT THE INVENTION

[0023] Embodiments of the present invention will be hereinafter described with reference to the accompanying drawings.

[0024] A first embodiment of the present invention is an optical signal transmitter in which N FM batch conversion circuits, to which electric signals distributed by a distribution circuit are input while being modulated, are used, and is an optical signal transmission system using this optical signal transmitter. This embodiment of the present invention is shown in Fig. 5. Fig. 5 shows a case in which N=3. In Fig. 5, the optical signal transmitter includes a distribution circuit 11, FM batch conversion

circuits 12, an optical multiplexing circuit 13, a light source 14 serving as a transmitting circuit, an optical amplification circuit 15, and an optical transmission path 85. The light source 14 may include a semiconductor laser 5 and a drive circuit that drives this semiconductor laser as a transmitting circuit, and the transmitting circuit may include the optical amplification circuit 15.

[0025] In Fig. 5, when multichannel AM video signals or QAM video signals that have undergone frequency multiplication in the frequency range of about 90 MHz to about 750 MHz as shown in Fig. 2A are input to the optical signal transmitter 10, the signals are distributed into three groups of signals by the distribution circuit 11. Each output of the distribution circuit 11 is input to the 15 FM batch conversion circuits 12 as a modulated input, and is subjected to frequency modulation by the FM batch conversion circuits 12. Outputs of the three FM batch conversion circuits 12 are multiplexed by the optical multiplexing circuit 13. The output of the optical multiplexing circuit 13 is a wideband frequency-modulated 20 electric signal as shown in Fig. 2B. This frequency-modulated electric signal is converted into an optical signal subjected to intensity modulation by the light source 14. The optical signal is amplified to a predetermined optical level by the optical amplification 25 circuit 15, and is transmitted to the optical transmission path 85.

[0026] Herein, if the three FM batch conversion circuits 12 are set to be equal to each other in frequency deviation and in intermediate frequency and are set to be identical to each other in the phase of each output of FM batch conversion circuits, the electric signals multiplexed by the optical multiplexing circuit 13 have their noise quantities expressed as the sum total of electric powers of the three FM batch conversion circuits 12, i.e., as an electric-power addition, and have their signal components expressed as the sum total of voltages thereof, i.e., as a voltage addition. Since the three FM batch conversion circuits 12 are set to be identical to each other in the phase of each output, it is possible to, for example, adjust the length of a transmission path, such as an optical fiber, 15 or use a phase adjuster.

[0027] Let the voltages of signal components output from the three FM batch conversion circuits 12 be designated as  $Vs1$ ,  $Vs2$ , and  $Vs3$ , respectively, and let  $Vs1=Vs2=Vs3=Vs$ . In this case, the sum total  $Vst$  of the voltages of signal components output from the optical multiplexing circuit 13 are expressed as follows:

$$Vst=Vs1+Vs2+Vs3=3Vs \quad (3)$$

[0028] Under the condition that the output of only one of the three FM batch conversion circuits 12 is input to the optical multiplexing circuit 13, the signal power  $Ps1$  25 of the output of the optical multiplexing circuit 13 is expressed as follows:

$$Ps1=Vs^2/R \quad (4)$$

where R is an input impedance of the light source 14. Under the condition that the outputs of the three FM batch conversion circuits 12 are input to the optical multiplexing circuit 13, the signal power  $Pst$  of the output of the optical multiplexing circuit 13 is expressed as follows:

$$Pst=(Vst)^2/R=9Vs^2/R \quad (5)$$

Therefore, the electric-power ratio between the signal power  $Ps1$  and the signal power  $Ps3$  is expressed as follows:

$$10\log(Pst/Ps1)=20\log(3)[\text{dB}] \quad (6)$$

[0029] On the other hand, let the electric powers of noise components output from the three FM batch conversion circuits 12 be designated as  $Pn1$ ,  $Pn2$ , and  $Pn3$ , respectively, and let  $Pn1=Pn2=Pn3=Pn$ . Since an electric-power addition is applied to noise components, the sum total  $Pnt$  of the electric powers of noise components output from the optical multiplexing circuit 13 is expressed as follows:

$$Pnt=Pn1+Pn2+Pn3=3Pn \quad (7)$$

If the output of only one of the three FM batch conversion circuits 12 is input to the optical multiplexing circuit 13, the noise power  $Pn1$  output from the optical multiplexing circuit 13 is expressed as follows:

$$Pn1=Pn \quad (8)$$

Therefore, the electric-power ratio between the noise power  $Pn1$  and the noise power  $Pnt$  is expressed as follows:

$$10\log(Pnt/Pn1)=10\log(3)[\text{dB}] \quad (9)$$

[0030] From this fact, it is understood that, in a case

in which the three FM batch conversion circuits are used, the signal power ratio becomes equal to  $20\log(3)[\text{dB}]$ , but the noise power ratio becomes equal to  $10\log(3)[\text{dB}]$ , and hence the signal-to-noise power in the output of the optical 5 multiplexing circuit is improved by  $10\log(3)[\text{dB}]$  in comparison with a case in which only one of the three FM batch conversion circuits is used. Although the structure using the three FM batch conversion circuits is shown in the embodiment of Fig. 5, the signal-to-noise power can 10 be improved by using two or more FM batch conversion circuits. In a case in which  $N$  FM batch conversion circuits are used (N is an integer which is two or greater), the signal-to-noise power can be improved by  $10\log(N)[\text{dB}]$  in comparison with a case in which only one FM batch conversion circuit is 15 used.

[0031] With regard to distortions, the three FM batch conversion circuits are different from each other in distortion characteristics, and, if they have distortion characteristics opposite in direction, offsetting can be 20 achieved in proportion to opposite distortions by a wave combination, and hence the distortions can be made lower than a case in which only one FM batch conversion circuit is used.

[0032] If the optical signal transmitter 10 of Fig. 5, 25 instead of the optical transmitter 80, is applied to the optical signal transmission system in Fig. 1, the minimum optical signal light-receiving electric power of the optical signal

receiver can be reduced, and the transmission distance can be lengthened, and the optical branching ratio can be enlarged between the optical signal transmitter and the optical signal receiver. Additionally, if low distortion characteristics can be realized by the optical signal transmitter, the quality of receiving video signals can be improved.

[0033] Although the signal of Fig. 2A is used as an example of a signal to be input to the optical signal transmitter in this embodiment, the invention is not limited to this signal form.

[0034] Next, a second embodiment of the present invention is a structure of an FM batch conversion circuit that is applied to the optical signal transmitter described in the first embodiment and that uses an optical frequency modulation portion. This embodiment of the present invention is shown in Fig. 6. In Fig. 6, the FM batch conversion circuit 12 includes an optical frequency conversion circuit 12, an optical frequency local modulation portion 22, an optical frequency local oscillation portion 32, an optical multiplexing portion 23, and an optical detector 24.

[0035] In the FM batch conversion circuit 12, when the frequency-multiplexed video signals shown in Fig. 2A are subjected to frequency modulation by use of the carrier light source having an optical frequency  $f_0$  in the optical frequency modulation portion 22, the optical frequency  $f_{fmld}$  of an optical signal in the output of the optical

frequency modulation portion 22 is calculated from Equation (1) mentioned above where  $\delta f$  is a frequency deviation. In Equation (1), the modulated signal is a signal having a frequency  $f_s$ . ADFB-LD (Distributed Feed-Back Laser Diode) 5 can be used as the carrier light source of the optical frequency modulation portion 22.

[0036] In the optical frequency local oscillation portion 32, oscillation is performed by use of an oscillating light source having an optical frequency  $f_1$ , and the signal 10 is multiplexed with an optical signal emitted from the optical frequency modulation portion 22 by the optical multiplexer 23. The DFB-LD can be used as the oscillating light source of the optical frequency local oscillation portion 32. The two optical signals multiplexed by the 15 optical multiplexer 23 are subjected to heterodyne detection by the optical detector 23. A photodiode that functions as a heterodyne detector can be used as the optical detector. The frequency  $f$  of the electric signal subjected to heterodyne detection by the optical detector 24 is 20 calculated from Equation (2) mentioned above. In Equation (2), the modulated signal is a signal having a frequency  $f_s$ . Herein, if the optical frequency of the carrier light source of the optical frequency modulation portion 22 and the optical frequency of the oscillating light source of 25 the local oscillation portion 32 are caused to come close to each other, it is possible to obtain an electric signal in which frequency is modulated to have an intermediate

frequency  $f_i = f_o - f_1$  of several GHz and have a frequency deviation  $\delta f$  as shown in Fig. 2B.

[0037] Generally, the modulation by an input electric current allows the DFB-LD to have an optical frequency varied 5 in the range of several GHz in accordance with the input electric current, and hence a value of several GHz can be obtained as the frequency deviation  $\delta f$ . For example, multichannel AM video signals or QAM video signals that have undergone frequency multiplication so as to have a 10 frequency range of about 90 MHz to about 750 MHz can be converted by the FM batch conversion circuit into a frequency-modulated signal having a frequency band of about 6 GHz in which the intermediate frequency  $f_i = f_o - f_1$  becomes equal to about 3GHz as shown in Fig. 2B.

[0038] Further, each intermediate frequency  $f_i$ , which 15 is a frequency equal to a difference between the optical frequency of the carrier light source of the optical frequency modulation portion 22 and the optical frequency of the oscillating light source of the optical frequency local oscillation portion 32 used in N FM batch conversion 20 circuits, is set to be substantially equal in the N FM batch conversion circuits, and frequency modulation is performed with substantially the same frequency deviation centering 25 on this intermediate frequency. Further, the N FM batch conversion circuits are set to be substantially identical to each other in the phase of each output. Thus, the output of the optical multiplexing circuit 13 of Fig. 5 has its

noise quantity expressed as the sum total of electric powers, i.e., as an electric-power addition and has its signal component expressed as the sum total of voltages, i.e., as a voltage addition. For example, the length of a transmission path, such as an optical fiber, can be adjusted, or a phase adjuster can be used, in order to set them so that the phase of each output becomes mutually identical.

5 [0039] From this fact, it is understood that, when use is made of an optical signal transmitter that uses N sets 10 of optical frequency modulation portions and optical frequency local oscillation portions, the signal power ratio becomes  $20\log(N)$  [dB], but the noise power ratio becomes  $10\log(N)$  [dB], and hence the signal-to-noise power 15 in the output of the optical multiplexing circuit 13 of Fig. 5 is improved by  $10\log(N)$  [dB] in comparison with a case in which use is made of an optical signal transmitter that uses one set of an optical frequency modulation portion and an optical frequency local oscillation portion.

20 [0040] With regard to distortions, the N sets of optical frequency modulation portions are different from each other in distortion characteristics, and, if they have distortion characteristics opposite in direction, offsetting can be achieved in proportion to opposite distortions by a wave combination, and hence the distortions can be made lower 25 than a case in which only one FM batch conversion circuit is used.

[0041] If the thus formed N FM batch conversion circuits

are applied to an optical signal transmitter, the minimum light-receiving electric power of an optical signal receiver in an optical signal transmission system can be reduced, and the transmission distance can be lengthened, 5 and the optical branching ratio can be enlarged between the optical signal transmitter and the optical signal receiver. Additionally, if low distortion characteristics can be obtained by the optical signal transmitter, the quality of receiving video signals can 10 be improved.

[0042] Although the signal of Fig. 2A is used as an example of a signal to be input to the optical signal transmitter in this embodiment, the invention is not limited to this signal form.

15 [0043] Next, a third embodiment of the present invention is a structure of an FM batch conversion circuit that is applied to the optical signal transmitter described in the first embodiment and that uses two optical frequency modulation portions for a push-pull structure. This 20 embodiment of the present invention is shown in Fig. 7. In Fig. 7, the FM batch conversion circuit 12 includes a differential distributor 21, an optical frequency modulation portion 22-1, an optical frequency modulation portion 22-2, an optical multiplexser 23, and an optical 25 portion 22-2, an optical multiplexser 23, and an optical detector 24.

[0044] In the FM batch conversion circuit 12, a frequency-multiplexed video signal, such as that shown in

Fig. 2A, is distributed by the differential distributor 21 into two electric signals whose phases have been inverted. If one of the two electric signals distributed by the differential distributor 21 is a modulated input, and if frequency modulation is performed by use of a carrier light source having an optical frequency  $f_{o1}$  in the optical frequency modulation portion 22-1, the optical frequency  $F_{fmld1}$  of an optical signal in the output of the optical frequency modulation portion 22-1 is expressed as follows:

$$F_{fmld1} = f_{o1} + (\delta f/2) \cdot \sin(2\pi \cdot f_s \cdot t) \quad (10)$$

where  $\delta f/2$  is a frequency deviation. In Equation (10), the modulated signal is a signal having a frequency  $f_s$ . If the other one of the two electric signals distributed by the differential distributor is a modulated input, and if frequency modulation is performed by use of a carrier light source having an optical frequency  $f_{o2}$  in the optical frequency modulation portion 22-2, the optical frequency  $F_{fmld2}$  of an optical signal in the output of the optical frequency modulation portion 22-2 is expressed as follows:

$$F_{fmld2} = f_{o2} - (\delta f/2) \cdot \sin(2\pi \cdot f_s \cdot t) \quad (11)$$

where  $\delta f/2$  is a frequency deviation. In Equation (11), the modulated signal is a signal having a frequency  $f_s$ . A DFB-LD (Distributed Feed-Back Laser Diode) can be used as a carrier light source for the optical frequency modulation portions 22-1 and 22-2.

[0045] Outputs emitted from the optical frequency modulation portions 22-1 and 22-2 are multiplexed by the

optical multiplexer 23, and the two optical signals multiplexed by the optical multiplexer 23 are subjected to heterodyne detection by the optical detector 23. A photodiode that functions as a heterodyne detector can be used as the optical detector. The frequency  $f$  of the electric signal subjected to heterodyne detection by the optical detector 24 is expressed as a frequency equal to a difference between the values shown in Equations (10) and (11) as follows:

$$f = f_{o1} - f_{o2} + \delta f \cdot \sin(2\pi \cdot f_s \cdot t) \quad (12)$$

In Equation (12), the modulated signal is a signal having a frequency  $f_s$ . Herein, if the optical frequency of the carrier light source of the optical frequency modulation portion 22-1 and the optical frequency of the carrier light source of the optical frequency modulation portion 22-2 are caused to come close to each other, it is possible to obtain an electric signal in which frequency is modulated to have an intermediate frequency  $f_i = f_o - f_1$  of several GHz and have a frequency deviation  $\delta f$  as shown in Fig. 2B.

[0046] Generally, the modulation by an input electric current allows the DFB-LD to have an optical frequency varied in the range of several GHz in accordance with the input electric current, and hence a value of several GHz can be obtained as the frequency deviation  $\delta f$ . For example, multichannel AM video signals or QAM video signals that have undergone frequency multiplication so as to have a frequency range of about 90 MHz to about 750 MHz can be

converted by the FM batch conversion circuit into a frequency-modulated signal having a frequency band of about 6 GHz in which the intermediate frequency  $f_i = f_0 - f_1$  becomes equal to about 3GHz as shown in Fig. 2B.

5 [0047] Further, each intermediate frequency  $f_i$ , which is a frequency equal to a difference between the optical frequency of the carrier light source of the optical frequency modulation portion 22-1 and the optical center frequency of the carrier light source of the optical frequency modulation portion 22-2 used in N FM batch conversion circuits, is set to be substantially equal in the N FM batch conversion circuits, and frequency modulation is performed with substantially the same frequency deviation centering on this intermediate frequency.

10 15 Further, the N FM batch conversion circuits are set to be substantially identical to each other in the phase of each output. Thus, the output of the optical multiplexing circuit 13 of Fig. 5 has its noise quantity expressed as the sum total of electric powers, i.e., as an electric-power addition and has its signal component expressed as the sum total of voltages, i.e., as a voltage addition. For example, the length of a transmission path, such as an optical fiber, can be adjusted, or a phase adjuster can be used, in order to set them so that the phase of each output becomes mutually identical.

20 25 [0048] From this fact, it is understood that, when use is made of an optical signal transmitter that uses N sets

of optical frequency modulation portions, the signal power ratio becomes  $20\log(N)$  [dB], however, the noise power ratio becomes  $10\log(N)$  [dB], and hence the signal-to-noise power in the output of the optical multiplexing circuit 13 of 5 Fig. 5 is improved by  $10\log(N)$  [dB] in comparison with a case in which use is made of an optical signal transmitter that uses one set of optical frequency modulation portions. [0049] With regard to distortions, the  $N$  sets of optical frequency modulation portions are different from each other 10 in distortion characteristics, and, if they have distortion characteristics opposite in direction, offsetting can be achieved in proportion to opposite distortions by a wave combination, and hence the distortions can be made lower than a case in which only one FM batch conversion circuit 15 is used.

[0050] If the thus formed  $N$  FM batch conversion circuits are applied to an optical signal transmitter, the minimum light-receiving electric power of an optical signal receiver in an optical signal transmission system can be 20 reduced, and the transmission distance can be lengthened, and the optical branching ratio can be enlarged between the optical signal transmitter and the optical signal receiver. Additionally, if low distortion characteristics can be obtained by the optical signal 25 transmitter, the quality of receiving video signals can be improved.

[0051] Although the signal of Fig. 2A is used as an

example of a signal to be input to the optical signal transmitter in this embodiment, the invention is not limited to this signal form.

[0052] Next, a fourth embodiment of the present invention is an FM batch conversion circuit that is applied to the optical signal transmitter described in the first embodiment and that uses a voltage-controlled oscillator. This embodiment of the present invention is shown in Fig. 8. In Fig. 8, the FM batch conversion circuit 12 includes a voltage-controlled oscillator 26.

[0053] In the FM batch conversion circuit 12, a frequency-multiplexed video signal, such as that shown in Fig. 2A, is subjected to frequency modulation with a frequency  $f_0$  as the center frequency in the voltage-controlled oscillator 26, and a frequency  $f_v$  of an electric signal that has been output is expressed as follows when the frequency deviation is  $\delta f$ :

$$f_v = f_0 + \delta f \cdot \sin(2\pi \cdot f_s \cdot t) \quad (13)$$

Thus, a frequency-modulated signal is obtained which has an intermediate frequency  $f_i = f_0$  and a frequency deviation  $\delta f$ . In Equation (13), the modulated signal is a signal having a frequency  $f_s$ .

[0054] For example, multichannel AM video signals or QAM video signals that have undergone frequency multiplication so as to have a frequency range of about 90 MHz to about 750 MHz can be converted by the FM batch conversion circuit into a frequency-modulated signal having

a frequency band of about 6 GHz in which the intermediate frequency  $f_i = f_o - f_1$  becomes equal to about 3 GHz as shown in Fig. 2B.

[0055] Further, each intermediate frequency  $f_i$  of the 5 voltage-controlled oscillator 26 used in N FM batch conversion circuits, is set to be substantially equal in the NFM batch conversion circuits, and frequency modulation is performed with substantially the same frequency deviation centering on this intermediate frequency. 10 Further, the N FM batch conversion circuits are set to be substantially identical to each other in the phase of each output. Thus, the output of the optical multiplexing circuit 13 of Fig. 5 has its noise quantity expressed as the sum total of electric powers, i.e., as an electric-power 15 addition and has its signal component expressed as the sum total of voltages, i.e., as a voltage addition. For example, the length of a transmission path, such as an optical fiber, can be adjusted, or a phase adjuster can be used, in order to set them so that the phase of each output becomes mutually 20 identical.

[0056] From this fact, it is understood that, when use is made of an optical signal transmitter that uses N voltage-controlled oscillators, the signal power ratio becomes  $20\log(N)$  [dB], however, the noise power ratio 25 becomes  $10\log(N)$  [dB], and hence the signal-to-noise power in the output of the optical multiplexing circuit 13 of Fig. 5 is improved by  $10\log(N)$  [dB] in comparison with a

case in which use is made of an optical signal transmitter that uses only one voltage-controlled oscillator.

[0057] With regard to distortions, the  $N$  voltage-controlled oscillators are different from each other in distortion characteristics, and, if they have distortion characteristics opposite in direction, offsetting can be achieved in proportion to opposite distortions by a wave combination, and hence the distortions can be made lower than a case in which only one FM batch conversion circuit is used.

[0058] If the thus formed  $N$  FM batch conversion circuits are applied to an optical signal transmitter, the minimum light-receiving electric power of an optical signal receiver in an optical signal transmission system can be reduced, and the transmission distance can be lengthened, and the optical branching ratio can be enlarged between the optical signal transmitter and the optical signal receiver. Additionally, if low distortion characteristics can be obtained by the optical signal transmitter, the quality of receiving video signals can be improved.

[0059] Although the signal of Fig. 2A is used as an example of a signal to be input to the optical signal transmitter in this embodiment, the invention is not limited to this signal form.

[0060] Next, a fifth embodiment of the present invention is an FM batch conversion circuit that is applied to the

optical signal transmitter described in the first embodiment and that uses two voltage-controlled oscillators for a push-pull structure. This embodiment of the present invention is shown in Fig. 9. In Fig. 9, the FM batch conversion circuit 12 includes a differential distributor 21, a voltage-controlled oscillator 28-1, a voltage-controlled oscillator 28-2, a mixer 29, and a low-pass filter 30.

[0061] In the FM batch conversion circuit 12, a frequency-multiplexed video signal, such as that shown in Fig. 2A, is distributed by the differential distributor 21 into two electric signals in which phases have been inverted. If one of the two electric signals distributed by the differential distributor 21 is subjected to frequency modulation with a frequency  $f_0$  as the center frequency in the voltage-controlled oscillator 28-1, a frequency  $f_{V1}$  of an electric signal output therefrom is expressed as follows:

$$f_{V1} = f_0 + (\delta f/2) \cdot \sin(2\pi \cdot f_s \cdot t) \quad (14)$$

where  $\delta f/2$  is a frequency deviation. Thus, a frequency-modulated signal which has an intermediate frequency  $f_i = f_0$  and a frequency deviation  $\delta f/2$  is obtained. In Equation (14), the modulated signal is a signal having a frequency  $f_s$ . If the other one of the two electric signals distributed by the differential distributor 21 is a modulated input, and is subjected to frequency modulation with a frequency  $f_0$  as the center frequency in the

voltage-controlled oscillator 28-2, a frequency  $f_{v2}$  of an electric signal output therefrom is expressed as follows:

$$f_{v2}=f_{o2} - (\delta f/2) \cdot \sin(2\pi \cdot f_s \cdot t) \quad (15)$$

where  $\delta f/2$  is a frequency deviation. Thus, a frequency-modulated signal which has an intermediate frequency  $f_i=f_{o2}$  and a frequency deviation  $\delta f/2$  is obtained. In Equation (15), the modulated signal is a signal having a frequency  $f_s$ .

[0062] Outputs from the voltage-controlled oscillators 28-1 and 28-2 are mixed together by the mixer 29. The two electric signals mixed together by the mixer 29 are then smoothed by the low-pass filter 30. The frequency  $f$  of the electric signal smoothed by the low-pass filter 30 that transmits an electric signal having a frequency equal to a difference between the intermediate frequency  $f_{o1}$  and the intermediate frequency  $f_{o2}$  is expressed as that of an electric signal having a frequency equal to a difference between the value of Equation (14) and the value of Equation (15) as follows:

$$f=f_{o1}-f_{o2}+\delta f \cdot \sin(2\pi \cdot f_s \cdot t) \quad (16)$$

In Equation (16), the modulated signal is a signal having a frequency  $f_s$ . Herein, it is possible to obtain an electric signal whose frequency is modulated to have an intermediate frequency  $f_i=f_{o1}-f_{o2}$  of several GHz and have a frequency deviation  $\delta f$  as shown in Fig. 2B.

[0063] For example, multichannel AM video signal or QAM video signal that have undergone frequency multiplication

so as to have a frequency range of about 90 MHz to about 750 MHz can be converted by the FM batch conversion circuit into a frequency-modulated signal having a frequency band of about 6 GHz in which the intermediate frequency  $f_i = f_o - f_1$  becomes equal to about 3GHz as shown in Fig. 2B.

[0064] Further, each intermediate frequency  $f_i$ , which is a frequency equal to a difference between the voltage-controlled oscillator 28-1 and the voltage-controlled oscillator 28-2 used in N FM batch conversion circuits, is set to be substantially equal in the NFM batch conversion circuits, and frequency modulation is performed with substantially the same frequency deviation centering on this intermediate frequency. Further, the N FM batch conversion circuits are set to be substantially identical to each other in the phase of each output. Thus, the output of the optical multiplexing circuit 13 of Fig. 5 has its noise quantity expressed as the sum total of electric powers, i.e., as an electric-power addition and has its signal component expressed as the sum total of voltages, i.e., as a voltage addition. For example, the length of a transmission path, such as an optical fiber, can be adjusted, or a phase adjuster can be used, in order to set them so that the phase of each output becomes mutually identical.

[0065] From this fact, it is understood that, when use is made of an optical signal transmitter that uses N voltage-controlled oscillators, the signal power ratio

becomes  $20\log(N)$  [dB], but the noise power ratio becomes  $10\log(N)$  [dB], and hence the signal-to-noise power in the output of the optical multiplexing circuit 13 of Fig. 5 is improved by  $10\log(N)$  [dB] in comparison with a case in which use is made of an optical signal transmitter that uses only one voltage-controlled oscillator.

[0066] With regard to distortions, the  $N$  optical frequency modulation portions are different from each other in distortion characteristics, and, if they have distortion characteristics opposite in direction, offsetting can be achieved in proportion to opposite distortions by a wave combination, and hence the distortions can be made lower than a case in which only one FM batch conversion circuit is used.

[0067] If the thus formed  $N$  FM batch conversion circuits are applied to an optical signal transmitter, the minimum light-receiving electric power of an optical signal receiver in an optical signal transmission system can be reduced, and the transmission distance can be lengthened, and the optical branching ratio can be enlarged between the optical signal transmitter and the optical signal receiver. Additionally, if low distortion characteristics can be obtained by the optical signal transmitter, the quality of receiving video signals can be improved.

[0068] Although the signal of Fig. 2A is used as an example of a signal to be input to the optical signal

transmitter in this embodiment, the invention is not limited to this signal form.

[0069] Next, a sixth embodiment of the present invention is an optical signal transmitter using two sets of optical frequency modulation portions and optical frequency local oscillation portions, and is an optical signal transmission system using this optical signal transmitter. This embodiment of the present invention is shown in Fig. 10. In Fig. 10, the optical signal transmitter 10 includes a distribution circuit 11, an optical frequency modulation portion 22-1, an optical frequency modulation portion 22-2, an optical frequency local oscillation portion 32-1, an optical frequency local oscillation portion 32-2, an optical multiplexer 25-1, an optical multiplexer 25-2, an optical multiplexer 27, an optical detector 24, a light source 14 serving as a transmitting circuit, an optical amplification circuit 15, and an optical transmission path 85. The light source 14 may include a semiconductor laser and a drive circuit, which drives this semiconductor laser, serving as a transmitting circuit. The transmitting circuit may include the optical amplification circuit 15.

[0070] In Fig. 10, when multichannel AM video signals or QAM video signals that have undergone frequency multiplication so as to have a frequency range of about 90 MHz to about 750 MHz as shown in Fig. 2A are input to the optical signal transmitter 10, the signals are distributed by the distribution circuit 11 into two signal

parts. One output of the distribution circuit 11 is input to the optical frequency modulation portion 22-1 as a modulated input, and is subjected to frequency modulation. The other output of the distribution circuit 11 is input 5 to the optical frequency modulation portion 22-2 as a modulated input, and is subjected to frequency modulation.

[0071] An optical signal subjected to frequency modulation by the optical frequency modulation portion 22-1 is multiplexed with local oscillation light emitted from 10 the local oscillation portion 32-1 by the optical multiplexer 25-1 while being caused to have the same polarization direction. Herein, the optical frequency of the optical frequency local oscillation portion 32-1 is apart from the optical center frequency of the 15 frequency-modulated optical signal output from the optical frequency modulation portion 22-1 by a frequency substantially equal to the intermediate frequency.

[0072] An optical signal subjected to frequency modulation by the optical frequency modulation portion 22-2 is multiplexed with local oscillation light emitted from 20 the local oscillation portion 32-2 by the optical multiplexer 25-2 while being caused to have the same polarization direction. Herein, the optical frequency of the optical frequency local oscillation portion 32-2 is apart from the optical center frequency of the 25 frequency-modulated optical signal output from the optical frequency modulation portion 22-2 by a frequency

substantially equal to the intermediate frequency.

[0073] The optical signals output from the optical multiplexers 25-1 and 25-2, respectively, are multiplexed by the optical multiplexer 27 while making the polarization 5 direction of the optical signal output from the optical multiplexer 25-1 perpendicular to the polarization direction of the second optical signal output from the optical multiplexer 25-2, and a multiplexed signal is output therefrom. The optical signal output from the optical 10 multiplexer 27 is subjected to heterodyne detection in the optical detector 24, and an electric signal, which has a frequency equal to a difference between the optical frequency of the optical signal emitted from the optical frequency modulation portion and the optical frequency of 15 the local oscillation light emitted from the local oscillation portion, is output. A photodiode that performs heterodyne detection can be used as the detector 24. The output of this detector 24 is a wideband frequency-modulated 20 electric signal as shown in Fig. 2B. This frequency-modulated electric signal is converted into an optical signal subjected to intensity modulation by the light source 14, is then amplified to a predetermined optical level by the optical amplification circuit 15, and is transmitted to the optical transmission path 85. A 25 semiconductor laser, such as a DFB-LD, can be used as the light source.

[0074] Herein, the frequency deviations of the two

optical frequency modulation portions 22-1 and 22-2 are set to be substantially equal to each other. Further, a difference between the optical frequency of the optical signal of the optical frequency modulation portion 22-1 and the optical frequency of the local oscillation light of the local oscillation portion 32-1 is set to be substantially equal to a difference between the optical frequency of the optical signal of the optical frequency modulation portion 22-2 and the optical frequency of the local oscillation light of the local oscillation portion 32-2. Further, the phase of an electric signal obtained by subjecting the multiplexed optical signal emitted from the optical multiplexer 25-1 to heterodyne detection by the optical detector 24 is set to be substantially equal to the phase of an electric signal obtained by subjecting the multiplexed optical signal emitted from the optical multiplexer 25-2 to heterodyne detection by the optical detector 24. Thereby, the electric signal detected by the optical detector 24 has its noise quantity expressed as the sum total of electric powers, i.e., as an electric-power addition and has its signal component expressed as the sum total of voltages, i.e., as a voltage addition. For example, the length of a transmission path, such as an optical fiber, can be adjusted, or a phase adjuster can be used, in order to set them so that the phase of each output becomes mutually identical.

[0075] From this fact, it is understood that, when use

is made of an optical signal transmitter that uses two sets of optical frequency modulation portions and optical frequency local oscillation portions, the signal power ratio becomes  $20\log(2)$  [dB], however, the noise power ratio 5 becomes  $10\log(2)$  [dB], and hence the signal-to-noise power in the output of the optical multiplexing circuit is improved by  $10\log(2)$  [dB] in comparison with a case in which use is made of an optical signal transmitter that uses only one set of an optical frequency modulation portion and an optical 10 frequency local oscillation portion.

[0076] With regard to distortions, the two sets of optical frequency modulation portions are different from each other in distortion characteristics, and, if they have distortion characteristics opposite in direction, offsetting can be achieved in proportion to opposite 15 distortions by a wave combination, and hence the distortions can be made lower than a case in which only one FM batch conversion circuit is used.

[0077] If the optical signal transmitter 10 of Fig. 10, 20 instead of the optical transmitter 80, is applied to the optical signal transmission system in Fig. 1, the minimum light-receiving electric power of the optical signal receiver can be reduced, and the transmission distance can be lengthened, and the optical branching ratio can be 25 enlarged between the optical signal transmitter and the optical signal receiver.

[0078] Additionally, if low distortion characteristics

can be realized by the optical signal transmitter, the quality of receiving video signals can be improved.

[0079] Although the signal of Fig. 2A is used as an example of a signal to be input to the optical signal transmitter in this embodiment, the invention is not limited to this signal form.

[0080] Next, a seventh embodiment of the present invention is an optical signal transmitter including two sets of two push-pull structure optical frequency modulation portions therein, and is an optical signal transmission system using this optical signal transmitter. This embodiment of the present invention is shown in Fig. 11. In Fig. 11, the optical signal transmitter 10 comprises a distribution circuit 11, a differential distributor 21-1, a differential distributor 21-2, an optical frequency modulation portion 22-1, an optical frequency modulation portion 22-2, an optical frequency modulation portion 22-3, an optical frequency modulation portion 22-4, an optical multiplexer 25-1, an optical multiplexer 25-2, an optical multiplexer 27, an optical detector 24, a light source 14 serving as a transmitting circuit, an optical amplification circuit 15, and an optical transmission path 85. The light source 14 may include a semiconductor laser and a drive circuit that drives this semiconductor laser as a transmitting circuit, and the transmitting circuit may include the optical amplification circuit 15.

[0081] In Fig. 11, when multichannel AM video signals

or QAM video signals that have undergone frequency multiplication so as to have a frequency range of about 90 MHz to about 750 MHz as shown in Fig. 2A are input to the optical signal transmitter 10, the signals are distributed by the distribution circuit 11 into two signal parts. One output of the distribution circuit 11 is distributed by the differential distributor 21-1 into two electric signals in which phases have been inverted. The optical frequency  $F_{fml1}$  of the output light emitted from the optical frequency modulation portion 22-1 is subjected to frequency modulation by one of the two electric signals emitted from the differential distributor 21-1, and thereby a frequency-modulated optical signal is output. The optical frequency  $F_{fml2}$  of the output light emitted from the optical frequency modulation portion 22-2 is subjected to frequency modulation by the other one of the two electric signals emitted from the differential distributor 21-1, and thereby a frequency-modulated optical signal is output. The frequency-modulated optical signal emitted from the optical frequency modulation portion 22-1 and the frequency-modulated optical signal emitted from the optical frequency modulation portion 22-2 are set so that a difference in the optical center frequency becomes substantially equal to an intermediate frequency and so that coincidence of the polarization direction is achieved, and are then multiplexed by the optical multiplexer 25-1, and are turned into a first optical signal.

[0082] The other output of the distribution circuit 11 is distributed by the differential distributor 21-2 into two electric signals in which phases have been inverted. The optical frequency Ffmlld3 of the output light emitted 5 from the optical frequency modulation portion 22-3 is subjected to frequency modulation by one of the two electric signals emitted from the differential distributor 21-2, and thereby a frequency-modulated optical signal is output. The optical frequency Ffmlld4 of the output light emitted 10 from the optical frequency modulation portion 22-2 is subjected to frequency modulation by the other one of the two electric signals emitted from the differential distributor 21-4, and thereby a frequency-modulated optical signal is output. The frequency-modulated optical signal 15 emitted from the optical frequency modulation portion 22-3 and the frequency-modulated optical signal emitted from the optical frequency modulation portion 22-4 are set so that a difference in the optical center frequency becomes substantially equal to an intermediate frequency and so that coincidence of the polarization direction is achieved, 20 are then multiplexed by the optical multiplexer 25-2, and are turned into a second optical signal.

[0083] The first optical signal output from the optical multiplexers 25-1 and the second optical signal output from 25 the optical multiplexers 25-2, are multiplexed by the optical multiplexer 27 while making the polarization direction of the first optical signal perpendicular to the

polarization direction of the second optical signal respectively, and a multiplexed signal is output. The optical signal output from the optical multiplexer 27 is subjected to heterodyne detection in the optical detector 5 24, and an electric signal is output. This electric signal has a frequency equal to a difference between the optical frequency of the frequency-modulated optical signal emitted from the optical frequency modulation portion 22-1 and the optical frequency of the frequency-modulated optical signal 10 emitted from the optical frequency modulation portion 22-2 and equal to a difference between the optical frequency of the frequency-modulated optical signal emitted from the modulation portion optical frequency 22-3 and the optical frequency of the frequency-modulated optical signal emitted 15 from the optical frequency modulation portion 22-4. A photodiode that performs heterodyne detection can be used as the detector 24. The output of this detector 24 is a wideband frequency-modulated electric signal as shown in Fig. 2B. This frequency-modulated electric signal is 20 converted into an optical signal subjected to intensity modulation by the light source 14, is then amplified to a predetermined optical level by the optical amplification circuit 15, and is transmitted to the optical transmission path 85. A semiconductor laser, such as a DFB-LD, can be 25 used as the light source.

[0084] Herein, the frequency deviations of the optical frequency modulation portion 22-1, the optical frequency

modulation portion 22-2 the optical frequency modulation portion 22-3, and the optical frequency modulation portion 22-4 are set to be substantially equal to each other. Also, the phase of an electric signal obtained by subjecting the 5 multiplexed optical signal emitted from the optical multiplexer 25-1 to heterodyne detection by the optical detector 24 is set to be substantially equal to the phase of an electric signal obtained by subjecting the multiplexed optical signal emitted from the optical multiplexer 25-2 to heterodyne detection by the optical detector 24. Thereby, 10 the electric signal detected by the optical detector 24 has its noise quantity expressed as the sum total of electric powers, i.e., as an electric-power addition and has its signal component expressed as the sum total of voltages, 15 i.e., as a voltage addition. For example, the length of a transmission path, such as an optical fiber, can be adjusted, or a phase adjuster can be used, in order to set them so that the phase of each output becomes mutually identical.

[0085] From this fact, it is understood that, when use 20 is made of an optical signal transmitter that uses two sets of optical frequency modulation portions, in which each set consisting of two optical frequency modulation portions serves as a push-pull structure, the signal power ratio becomes  $20\log(2)[\text{dB}]$ , however, the noise power ratio 25 becomes  $10\log(2)[\text{dB}]$ , and hence the signal-to-noise power in the output of the optical multiplexing circuit is improved by  $10\log(2)[\text{dB}]$  in comparison with a case in which use is

made of an optical signal transmitter that uses only one set of two optical frequency modulation portions serving as a push-pull structure.

[0086] With regard to distortions, the two optical 5 frequency modulation portions serving as a push-pull structure are different from each other in distortion characteristics, and, if they have distortion characteristics opposite in direction, offsetting can be achieved in proportion to opposite distortions by a wave 10 combination, and hence the distortions can be made lower than a case in which only one FM batch conversion circuit is used.

[0087] If the optical signal transmitter 10 of Fig. 11, instead of the optical transmitter 80, is applied to the 15 optical signal transmission system in Fig. 1, the minimum light-receiving electric power of the optical signal receiver can be reduced, and the transmission distance can be lengthened, and the optical branching ratio can be enlarged between the optical signal transmitter and the 20 optical signal receiver.

[0088] Additionally, if low distortion characteristics can be realized by the optical signal transmitter, the quality of receiving video signals can be improved.

[0089] Although the signal of Fig. 2A is used as an 25 example of a signal to be input to the optical signal transmitter in this embodiment, the invention is not limited to this signal form.

[0090] Next, an eighth embodiment of the present invention is an optical signal transmitter that uses N optical frequency modulation multiplexing circuits to which electric signals distributed by a distribution circuit are 5 input while being modulated, and is an optical signal transmission system using this optical signal transmitter. This embodiment of the present invention is shown in Fig. 12. In Fig. 12, the optical signal transmitter 10 comprises a distribution circuit 11, an optical frequency modulation 10 multiplexing circuit 33, an optical multiplexing circuit 34, an optical detection circuit 35, a light source 14 serving as a transmitting circuit, an optical amplification circuit 15, and an optical transmission path 85. The light source 14 may include a semiconductor laser and a drive circuit 15 that drives this semiconductor laser as a transmitting circuit, and the transmitting circuit may include the optical amplification circuit 15. A structure of the optical frequency modulation multiplexing circuit 33 is shown in Fig. 13. In Fig. 13, the optical frequency 20 modulation multiplexing circuit 33 comprises an optical frequency modulation portion 22, an optical frequency local oscillation portion 32, and an optical multiplexer 23.

[0091] In Fig. 12, when multichannel AM video signals or QAM video signals that have undergone frequency 25 multiplication so as to have a frequency range of about 90 MHz to about 750 MHz as shown in Fig. 2A are input to the optical signal transmitter 10, the signals are

distributed by the distribution circuit 11 into N signal parts. Fig. 12 shows a case in which N=3. The output of the distribution circuit 11 is input to each of the N optical frequency modulation multiplexing circuits 33 as a 5 modulated input, and is subjected to frequency modulation by the optical frequency modulation portion 22 shown in Fig. 13.

[0092] In the optical frequency modulation multiplexing circuit 33 shown in Fig. 13, the optical frequency modulation portion 22 outputs a frequency-modulated optical signal, and the optical frequency local oscillation portion 32 outputs an optical local oscillation signal having a frequency apart from the optical frequency of the optical signal output from the optical frequency modulation portion 15 22 by a frequency substantially equal to the intermediate frequency. The frequency-modulated optical signal and the output from the optical frequency local oscillation portion 32 are multiplexed by the optical multiplexer 23.

[0093] The multiplexed optical signals output from the 20 three optical frequency modulation multiplexing circuits 33 are multiplexed by the optical multiplexing circuit 34, are then subjected to heterodyne detection by the optical detection circuit 35, and are turned into an electric signal having a frequency equal to a difference between the optical 25 frequency of the frequency-modulated optical signal emitted from the optical frequency modulation portion and the optical frequency of the optical local oscillation signal

emitted from the optical frequency local oscillation portion. A photodiode can be used as the optical detection circuit 35. The output of the optical detection circuit 35 is a wideband frequency-modulated electric signal as 5 shown in Fig. 2B. This frequency-modulated electric signal is converted into an optical signal subjected to intensity modulation by the light source 14, is then amplified to a predetermined optical level by the optical amplification circuit 15, and is transmitted to the optical transmission 10 path 85. A semiconductor laser, such as a DFB-LD, can be used as the light source.

[0094] Herein, the frequency deviations of the  $N$  optical frequency modulation multiplexing circuits are set to be substantially equal to each other. Further, the phases 15 of electric signals obtained by subjecting the optical signals emitted from the  $N$  optical frequency modulation multiplexing circuits 33 to heterodyne detection by the optical detection circuit 35 are set to be substantially equal to each other. Thereby, the electric signals detected 20 by the optical detection circuit 35 have a noise quantity expressed as the sum total of electric powers, i.e., as an electric-power addition and have a signal component expressed as the sum total of voltages, i.e., as a voltage addition. For example, the length of a transmission path, 25 such as an optical fiber, can be adjusted, or a phase adjuster can be used, in order to set them so that the phase of each output becomes mutually identical.

[0095] From this fact, it is understood that, when use is made of an optical signal transmitter that uses N optical frequency modulation multiplexing circuits, the signal power becomes  $20\log(N)$ , however, the noise power becomes 5  $10\log(N)$ , and hence the signal-to-noise power in the output of the optical multiplexing circuit is improved by  $10\log(N)$  [dB] in comparison with a case in which use is made of an optical signal transmitter that uses only one optical frequency modulation multiplexing circuit.

[0096] With regard to distortions, the N optical frequency modulation portions are different from each other in distortion characteristics, and, if they have distortion characteristics opposite in direction, offsetting can be achieved in proportion to opposite distortions by a wave 15 combination, and hence the distortions can be reduced.

[0097] If the optical signal transmitter 10 of Fig. 12, instead of the optical transmitter 80, is applied to the optical signal transmission system in Fig. 1, the minimum light-receiving electric power of the optical signal 20 receiver can be reduced, and the transmission distance can be lengthened, and the optical branching ratio can be enlarged between the optical signal transmitter and the optical signal receiver. Additionally, if low distortion characteristics can be realized by the optical signal 25 transmitter, the quality of receiving video signals can be improved.

[0098] Although the signal of Fig. 2A is used as an

example of a signal to be input to the optical signal transmitter in this embodiment, the invention is not limited to this signal form.

[0099] Next, a ninth embodiment of the present invention 5 is an optical signal transmitter that uses N differential optical frequency modulation multiplexing circuits to which electric signals distributed by a distribution circuit are input while being modulated, and is an optical signal transmission system using this optical signal transmitter. 10 This embodiment of the present invention is shown in Fig. 14. Fig. 14 shows a case in which N=3. In Fig. 14, the optical signal transmitter 10 comprises a distribution circuit 11, a differential optical frequency modulation multiplexing circuit 36, an optical multiplexing circuit 15, and an optical detection circuit 35, a light source 14 serving 15 as a transmitting circuit, an optical amplification circuit 15, and an optical transmission path 85. The light source 14 may include a semiconductor laser and a drive circuit that drives this semiconductor laser as a transmitting 20 circuit, and the transmitting circuit may include the optical amplification circuit 15. A structure of the differential optical frequency modulation multiplexing circuit 36 is shown in Fig. 15. In Fig. 15, the differential optical frequency modulation multiplexing circuit 36 25 comprises a differential distributor 21, an optical frequency modulation portion 22-1, an optical frequency modulation portion 22-2, and an optical multiplexer 23.

[0100] In Fig. 14, when multichannel AM video signals or QAM video signals that have undergone frequency multiplication so as to have a frequency range of about 90 MHz to about 750 MHz as shown in Fig. 2A are input to 5 the optical signal transmitter 10, the signals are distributed by the distribution circuit 11 into N signal parts. The output of the distribution circuit 11 is input to each of the N differential optical frequency modulation multiplexing circuits 36 as a modulated input.

10 [0101] In the differential optical frequency modulation multiplexing circuit 36 shown in Fig. 15, the output from the distribution circuit 11 is distributed by the differential distributor 21 into two electric signals in which phases have been inverted. The two electric signals 15 are turned into frequency-modulated optical signals in the optical frequency modulation portion 22-1 and the optical frequency modulation portion 22-2, respectively. The optical frequency of an optical signal output from the optical frequency modulation portion 22-1 and the optical 20 center frequency of an optical signal output from the optical frequency modulation portion 22-2 are apart from each other by the intermediate frequency. The frequency-modulated optical signals emitted from the optical frequency modulation portion 22-1 and the optical frequency 25 modulation portion 22-2 are multiplexed by the optical multiplexer 23, and are output to the optical multiplexing circuit 34 shown in Fig. 14. Herein, the intermediate

frequencies in the N differential optical frequency modulation multiplexing circuits 36 are set to be substantially equal to each other.

[0102] The optical signals output from the N differential optical frequency modulation multiplexing circuits 36 are multiplexed by the optical multiplexing circuit 34, are then subjected to heterodyne detection by the optical detection circuit 35, and are turned into an electric signal having a frequency equal to a difference between the optical frequency of the frequency-modulated optical signal emitted from the optical frequency modulation portion 22-1 and the optical frequency of the frequency-modulated optical signal emitted from the optical frequency modulation portion 22-2. A photodiode can be used as the optical detection circuit 35. The output of the optical detection circuit 35 is a wideband frequency-modulated electric signal as shown in Fig. 2B. This frequency-modulated electric signal is converted into an optical signal subjected to intensity modulation by the light source 14, is then amplified to a predetermined optical level by the optical amplification circuit 15, and is transmitted to the optical transmission path 85. A semiconductor laser, such as a DFB-LD, can be used as the light source.

[0103] Herein, the frequency deviations of the N differential optical frequency modulation multiplexing circuits are set to be substantially equal to each other.

Further, the phases of electric signals obtained by subjecting the optical signals emitted from the N differential optical frequency modulation multiplexing circuits 36 to heterodyne detection by the optical detection circuit 35 are set to be substantially equal to each other. Thereby, the electric signals detected by the optical detection circuit 35 have a noise quantity expressed as the sum total of electric powers, i.e., as an electric-power addition and have a signal component expressed as the sum total of voltages, i.e., as a voltage addition. For example, the length of a transmission path, such as an optical fiber, can be adjusted, or a phase adjuster can be used, in order to set them so that the phase of each output becomes mutually identical.

[0104] From this fact, it is understood that, when use is made of an optical signal transmitter that uses N differential optical frequency modulation multiplexing circuits, the signal power becomes  $20\log(N)$ , however, the noise power becomes  $10\log(N)$ , and hence the signal-to-noise power in the output of the optical multiplexing circuit is improved by  $10\log(N)$  [dB].

[0105] With regard to distortions, the  $2N$  optical frequency modulation portions are different from each other in distortion characteristics, and, if they have distortion characteristics opposite in direction, offsetting can be achieved in proportion to opposite distortions by a wave combination, and hence the distortions can be reduced.

[0106] If the optical signal transmitter 10 of Fig. 14, instead of the optical transmitter 80, is applied to the optical signal transmission system in Fig. 1, the minimum light-receiving electric power of the optical signal receiver can be reduced, and the transmission distance can be lengthened, and the optical branching ratio can be enlarged between the optical signal transmitter and the optical signal receiver. Additionally, if low distortion characteristics can be realized by the optical signal transmitter, the quality of receiving video signals can be improved.

[0107] Although the signal of Fig. 2A is used as an example of a signal to be input to the optical signal transmitter in this embodiment, the invention is not limited to this signal form.

[0108] Additionally, the optical transmitter and the optical transmission system of the present invention can be used in a case in which the network of the optical transmission path is a passive double star topology (PDS topology), as well as a single star topology (SS topology).